The Effects of Radiation Losses on the Measurement of Loss Tangent Using Microstrip Ring Resonators.

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Motivation

• The loss tangent of a moisture absorbing material is linked to the moisture content.

• This can be utilized to find the moisture content through loss tangent measurements.

• This work resulted from a measurement technique developed to measure the moisture content of *transformer press boards*.

• Our focus was on the dielectrics in sheet form used in transformers used in power distribution networks.

• If the moisture content (in the dielectric insulation) is measured, the transformer failure can be predicted and the reliability can be improved.

Our application





Microstrip line based techniques

• We consider microstrip line based techniques at microwave frequencies.

• At microwave frequencies, the losses due to the presence of moisture are significant.

Ring resonator technique



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• Experimental set up and measured data.



Multi-layer dielectric stack



• Complications arise due to the multi-layer nature of the problem.

Basis of calculations

$$\frac{1}{Q_u} = \frac{1}{Q_c} + \frac{1}{Q_r} + \frac{1}{Q_d}$$

$$Q_u = \frac{f_0}{(1 - |S_{21}|)B_{3dB}}$$

Q_c: Hammerstad equations.

 Q_r : Neglect?

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p_i: Schneider's filling factor

Is radiation significant? The contribution of this paper.

Radiation losses are usually ignored in measurements.

• In some techniques, the radiation losses with and without the test sample are assumed to be equal and balanced out.

Radiation loss

 Although difficult, analytical formulas can be obtained for the radiation from the ring structure. However, these techniques ignore the effects of the feeder lines and gaps.

 Measurement needs an anechoic chamber and are not suitable for low-cost dielectric measurements.

Simulation using HFSS

 In this paper, we use finite element based EM simulation to accurately obtain the radiation losses.

The solution gives
 Ē and *H* at every



Finding Qs

• Constituant Q factors $(Q_d, Q_r, \text{and } Q_c)$ and Q_u can be found since \tilde{E} and \tilde{H} are known.

$$Q = \frac{w_m + w_e}{P_l} \cdot \omega$$

 w_e, w_m : Average electrical/magnetic stored energy.

 P_l : Energy dissipation per second.

Stored energy densities in terms of field components:

$$w_m = \frac{1}{2} Re \int \left(\widetilde{B} \cdot \widetilde{H}^*\right) dV$$

$$w_e = \frac{1}{2} Re \int \left(\widetilde{D} \cdot \widetilde{E}^* \right) dV$$

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Energy densities in terms of field components:

$$P_{rad} = \frac{1}{2} Re \int \left(\widetilde{E} \times \widetilde{H}^* \right) dS$$

$$P_c = \frac{1}{2} Re \int (\tilde{J}. \tilde{E}^*) dV$$

$$P_d = \frac{\omega}{2} \operatorname{Re} \int \left(\widetilde{D} \cdot \widetilde{E}^* \right) \mathrm{d} V$$

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Evaluation of integrals: HFSS field calculator

- \$begin 'Named_Expression'
- Name('DieledtricLoss')
- Expression('+(Integrate(Volume(substrate), Volume_Loss_Density), Integrate(Volume(TestSample), Volume_Loss_Density))')
- NameOfExpression('Volume_Loss_Density')
- EnterVolume('substrate')
- Operation('VolumeValue')
- Operation('Integrate')
- NameOfExpression('Volume_Loss_Density')
- EnterVolume('TestSample')
- Operation('VolumeValue')
- Operation('Integrate')
- Operation('+')
- \$end 'Named_Expression'

Results

Resonant frequency	4.037 GHz
3 dB Bandwidth	20 MHz
S21 at resonance	-23.96 dB
Conduction loss	8.5 mW
Dielectric loss	178.48 mW
Radiation loss	31.76 mW
Stored magnetic energy	$5.19 \times 10^{-7} \mathrm{mW}$
Stored electrical energy	$4.87 \times 10^{-7} \text{ mW}$

• Note: HFSS normalizes input power to 1W.

Results

Q_u from and eigen-solution with HFSS	115.2
Q_u from stored energy and loss (equation 12 in the paper)	116.2
Effective permittivity (equation 5 in the paper)	3.33
Relative permittivity of the test sample	5.18
Loss tangent with radiation accounted	0.03
Loss tangent with radiation ignored	0.04

• How do we know the above results are correct?

1. Measured vs. simulated results



2. Cross check the value of Q_u

- Since we know w_e , w_m , P_c , P_r , and P_d unloaded Q (Q_u) can be calculated.
- Alternatively, Q_u can be obtained from an eigen solution.
- Method 1: $Q_u = 116.2$
- Method 2: $Q_u = 115.2$
- Error = 1.2 %

Conclusion

Results showed that the radiation, dielectric, and conductor losses amount to 3.9%, 81.6%, 14.5% in the simulated problem.

• Results further showed that ignoring the radiation loss resulted in a 33.3% error.

On going work

• The above conclusions lead to focus our attention to strip-line based measurements which are free from radiation effects.

 The work presented is a part of a more detailed work which has been accepted for publication in August 2016, it IEEE transactions of Dielectrics and Insulators. Thank you.